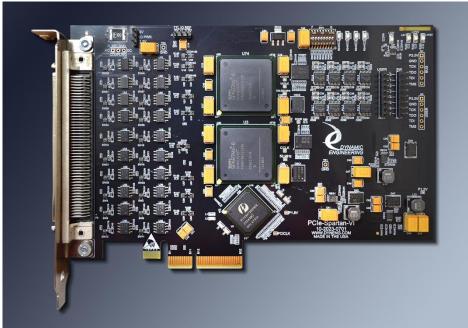
# DYNAMIC ENGINEERING

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**User Manual** 

# **PCIe-Spartan-VI**

# Re-configurable Logic with RS-485/LVDS and TTL IO



#### PCIe-Spartan-VI Rev 01

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**Embedded Solutions** 

#### PCIe-Spartan-VI

Re-configurable Logic PCIe Module Dynamic Engineering 150 DuBois St. Suite B&C, Santa Cruz CA 95060 831-457-8891

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## **Product Description**

PCIe-Spartan-VI provides a user configurable Spartan VI FPGA (XC6SLX100-3FGG676I), along with 40 RS-485 or 40 LVDS transceivers, 12 TTL IO, 24 PLL based clock references and FIFO support, full DMA capabilities in a half-length single slot design.

In some applications it is advantageous to adapt to the situation with updated parameters or even architectures. Other times it can be an advantage to load the program to use that day or for that situation. For example: in signal processing it may be necessary to change a filter – parameters can likely be handled with updated register values, but adding stages, changing types, perhaps changing compression are harder to do with parameters and not changing the underlying design. Since multiple design implementations can be stored and loaded on demand, the data can be tested and the appropriate filter or handling mechanism installed to operate with that data set. The load time is short and can be done with a few simple commands. See the JTAG programming file included with the Windows package.

PCIe-Spartan-VI provides two methods to program the Client design into the User FPGA. The programming header position connects to a standard programming interface to load the FPGA or FLASH with the program. The second method is to use software to program the FPGA or FLASH. The JTAG interface used by Impact and the Xilinx programming adapter is mimicked by the Bus FPGA, software and a mux to select between the options. The UserAp has several examples of "programming" the User configuration. Reading the Spartan VI ID, Loading a BIT file into the User FPGA, Programming file by using the record function in Impact. Please see the SW manual for more information on this process.

The RS-485 and LVDS parts can be mixed. The standard RS-485 devices are rated for 50 MHz. The standard LVDS parts are rated for 200 MHz.

The PCIe bus implementation incorporates 4 lanes using a [Pericom] PI7C9X130DNDE bridge. The bridge is connected through the Bus Controller FPGA to the User Spartan VI FPGA. The Bus Controller provides a GPB [General Purpose Bus] which operates as a 32 bit, 50 MHz connection for programming registers and other set-up / status operation. In addition, there are 16 unidirectional data lanes arranged as 8 bidirectional pairs to support transmit and receive operations between the User and Controller FPGAs. These are intended to be the main data paths. Each lane is supported with an independent DMA controller.

The Bus Controller supports multiple channels of data-flow to the User FPGA with FIFOs. There are 8 transmit and 8 receive FIFO paths. The "transmit" path is loaded using DMA transfers to read data from the host into the Bus Controller memory. Local



transfer engines move the data to the User FPGA side using FIFO status to control the transfer. With 8 DMA controllers each data path is separate and can operate in parallel. The "receive" path works in reverse with transfer engines in the User FPGA moving data into the Bus Controller using the FIFO status to control the transfer. Data is DMA transferred from the local FIFO within the Bus Controller to host memory. 8 DMA engines are available to manage the receive direction allowing for 8 ports of full duplex operation. Within the Bus Controller is a local arbitrator to control access to the intermediate bus. The controller is automatic – no software required.

The data is moved between the two FPGAs with a "push", the target side provides a FULL and Almost Full status to the data mover. The data transfer engine pushes data from the local FIFO to the target FIFO. When both the local and target FIFOs are not almost empty and not almost full respectively; the data is moved in a burst. 1 byte per clock. Due to pipelining, there are delays from write to current status being available to the control engine. When Master and or Target is "almost" (Full or Empty) the data transfer engine slows down to get exact status.

PCIe-Spartan-VI has drivers for Windows, and Linux. Both come with a sample user application to demonstrate the use of the various HW capabilities. The VHDL for the User FPGA reference design is also available. The reference design can be modified to perform your unique function. The reference design has 8 ports plus a base function. Built in support for internal and external loop-back of the data lanes is provided. The external loop-back [RS485 or LVDS] is in the form of a nibble wide data port with direction control[internal] and strobe. Your state machine can be substituted for the IO controller in the reference design.

There are 8 channels in the reference design. If you need more IO per channel and fewer channels or some other mix the VHDL can be modified. The GPB is supported with IOCTL style support in the driver allowing for user redefinition of the memory map within the User FPGA.

The reference software demonstrates DMA operation through the external IO and internal BIT tests. In addition, non-DMA versions are provided.

The base reference design has a controller for programming the PLLs. The reference software shows how to program the interface. Automatic software for taking the Cypress .JED file, parsing and loading the PLL.

The base reference design also has a simple register based, interface for the TTL IO.

An 8-bit "dip switch" is provided on the PCIe-Spartan-VI-485/LVDS. The switch configuration is readable via a register. The switch is for user-defined purposes. We envision the switch being used for software configuration control, PCIe board identification or test purposes.



LEDs are provided on board for user-defined purposes. The User FPGA has 4 LEDs, which can be used for whatever purpose the user desires. For example: use the LEDs for test / debugging purposes – indications of what the hardware state is or what the current software process is. The LEDs in the reference design default to a non-zero pattern to show the FLASH file has loaded properly. The register controlling the LEDs can overwrite the default.

In addition LEDs are provided to indicate that the regulators are operating properly

The UserAp reference software can load a User FPGA design file into the User FPGA at any time. UserAp also contains a function to reload from FLASH to allow the user to toggle back and forth between design versions rapidly. User Designs can also be programmed into FLASH with the programming tool [Impact in this case].

The User FPGA will be loaded on power transitions automatically from FLASH. Software can overwrite.



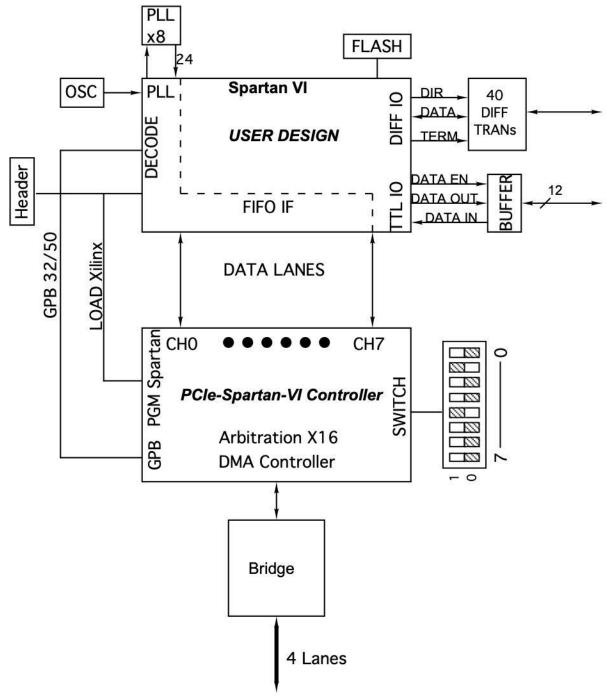


Figure 1

PCIe-Spartan-VI Block diagram

PCIe-Spartan-VI-485/LVDS has both RS-485/LVDS and TTL IO supported by a D100 connector. There are 40 un-committed RS-485 or LVDS IO. The standard RS-485 device supports 50 MHz operation. The LVDS parts have much larger bandwidth, 200 MHz receive and 400+ MHz transmit. The transceivers are designed for multi-drop



LVDS operation. The SN65MLVD201D is the part in the current design.

The 485/LVDS lines are routed as differential pairs with matched lengths and impedance control. The lengths are matched from the connector edge to the ball on the User FPGA. The right-angle connector is used for the matched length calculations. With the vertical connector option, the lengths will be slightly off. The differential signals are supported with programmable terminations. The terminations can be programmed from the User FPGA to be active or open.

The TTL IO is supported with open drain drivers with pull-ups. The receive side is also buffered to protect the User FPGA device. The pull-ups are referenced to a shunt selectable 3.3V or 5V. Data Out, Data In, and Data Enable controls for each IO. If Data Enable is programed to switch with Data Out the device will act as open drain. If set to Enabled or Disabled and Data Out operated independently the interface will be push-pull with 24mA source and sink for higher bandwidth than the open drain configuration.

The reference design has a pin configuration file, which can be reused for your specific implementation. The reference design is written in VHDL. For initial development and potentially as part of your system the D100 cable and the HDEterm100 are recommended. The HDEterm100 serves as a breakout from the cable to screw terminal block. The HDEterm100 has matched length, differential routing and several termination options that can be installed. For more information on the HDEterm100 please visit the web page <a href="https://www.dyneng.com/HDEterm100.html">https://www.dyneng.com/HDEterm100.html</a>

Clocking options in the User FPGA are supported with 8 PLL devices. The PLLs are programmable with a frequency description file. The PLLs are programmed via the User FPGA. The reference design has an interface to the PLLs implemented along with software to load the PLLs with the requested frequency file.

The User FPGA is connected to the 50 MHz GPB clock and 133.33 MHz oscillator plus the PLLs. Locally generated frequencies or one of the clock inputs can be routed to the PLLs. The reference design divides the 133.33 to provide 66.666 to the PLLs. The references can be unique if desired. The PLLs are Cypress 22393 devices. Cypress has a utility available for calculating the frequency control words for the PLLs. The PLLs respond to one of two addresses [only one works]. Our test SW "discovers" the working address and then uses that address to program the PLL, read-back, and run frequency tests.



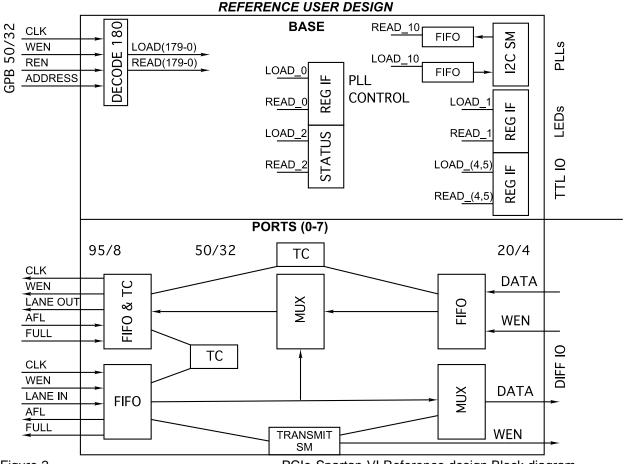


Figure 2

PCIe-Spartan-VI Reference design Block diagram

# Theory of Operation

A wide range of interfaces and protocols can be implemented with PCIe-Spartan-VI. UART, Manchester encoding, serial or parallel, RS-422/485 or TTL, custom. The interfaces can be created using the hardware and development tools provided with the PCIe-Spartan-VI along with the AMD/Xilinx software.

VHDL for the reference design is available. The reference design was implemented with ISE 14.7. The reference design is intended to be the starting point for a client design. The data lanes, GPB and other interfaces are implemented to allow the client to focus on the IO interface. The drivers for PCIe-Spartan-VI provide DMA and target accesses to the HW. The GPB interface is designed to use a passed in address rather than one hard coded in the driver so the user can change the memory map and not break the driver. The driver does put the address offset in so the client only needs to add the local [User] offset to create the pointer to the register to read or write. The reference SW has many examples and an include file with the offsets for our reference



application.

In the reference design, 8 channels are implemented each with a pair of FIFO data lanes attached. The design uses the 40 differential IO – 5 per channel. A simple nibble wide data transfer engine with strobe is implemented. The local 20 MHz clock [internal DCM] is used. The user design can remove the "end" code from the channel VHDL and replace with your state-machine etc. If you need different IO counts or Channel counts etc. a little editing will get you there. The code is commented, and the "find" command will help to navigate to the origin of signals etc. Nothing is "black boxed" so you can see what is being done and make changes as desired. While the VHDL is copyright to Dynamic Engineering, it is a purchase once and use many license. It is intended to be used, and reused by our clients. We want our clients to succeed, working to keep the path to success open.

Once your requirements are known the design can be implemented with VHDL, Verilog, or schematics and compiled with the design software. The output file can then be "uploaded" to the User FPGA on PCIe-Spartan-VI. Because the FPGA can be reloaded, your design can be implemented in phases. You can experiment and test out concepts and partial implementations during the design phase or perhaps simulate other hardware that needs to be implemented. The User FPGA can be loaded with a BIT file directly or indirectly by programming the QSPI with the new image.

As an example, consider a serial interface with 8 channels. PCIe-Spartan-VI has 40 differential IO. Enough IO for 8 full duplex channels with a clock reference and two controls per channel. PCIe-Spartan-VI has 16 FIFOs with 8 oriented for transmitting data and 8 for receiving data, plus 8 PLLs. Each channel can be supported completely and independently. The User design could be based on the reference design; in which case the PLL interface, basic FIFO interface, and bus decoding are taken care of in the User FPGA plus all of the functions provided by the Bus Controller. The designer would need to implement the IO interface.

The data flow for transmission is Host memory transferred into the Bus Controller TX FIFO via DMA transfers. The data is transferred from the Bus Controller holding FIFOs to the User FPGA Holding FIFOs with the transfer control state-machines. Each channel is separately supported. The User FPGA side reports Almost Full and Full to allow the transfer control to burst when the Bus Controller FIFO is not Almost Empty and the User FIFO is not almost full. 95 MHz x8 per channel in each direction. The user state machine would read the data from the FIFO on the User FPGA side and apply the user protocol before transmitting. On the receive side, the data will flow into the User FPGA, be processed to convert to a format suitable for storing, and written into the associated channels' Rx FIFO. The data would be pushed from the User to the Bus Controller's channel Rx FIFO using similar controls to the TX path. Almost Full and Full flags coming from the Bus Controller in this case. The DMA engine automatically senses the data in the channel FIFO and moves to the host memory.



Flow control is implemented throughout the data paths. Arbitration is supplied within the Bus Controller to handle the 16 DMA engines automatically. Scatter Gather DMA is supported for arbitrarily large data transfers. Large segments are allowed for OS with larger segment sizes [Linux compared to Windows®]. One interrupt per DMA at the end.

Other reference design features: The PLL i2c interface is provided with a state-machine rather than "bit banging". Two FIFOs are provided to allow a complete PLL programming file to be loaded and transmitted to the intended PLL. PLL programming data can be retrieved from the selected PLL and stored into the receive FIFO. ACK/NAK status can be used to determine the address for the PLL. Reference software in the "UserAp" code that comes with each driver type.

Besides removing the overhead of the CPU doing the bit banging an advantage of this approach is the PLL being programmed more rapidly. The i2c SM can send the bits at the maximum rate the PLL is rated for. It is difficult to manage this in SW resulting in longer than necessary programming times.

The decoder shown in the diagram interfaces with the GPB to generate READ and LOAD pulses. To implement a register, tie the LOAD to the clock enable, DataIn to the D inputs and the reference clock to the clock. To read-back, enable the corresponding mux path with the READ strobe. There are 180 of each type of strobe available. The reference design allocates 0-19 to the base design, and the rest to the channels. Channel 0 using 20-39. Channel 1 using 40-59 etc. 20 - 32 bit decodes per port is frequently enough.



#### Feature List Current

- User Defined Spartan VI series FPGA
- On-The-Fly programming of the User FPGA
- DMA and target accesses
- 16 FIFO based data lanes 8 dedicated "RX" and 8 dedicated "TX"
- 40 fully programmable RS485/422 or LVDS IO
- 12 TTL 10
- 8 PLLs each with 3 clocks supplied to Spartan VI
- 8 position "DIP Switch"
- User LEDs
- Power LEDs
- FLASH programming of User FPGA QSPI

As Dynamic Engineering adds features to the hardware, we will update the PCIe-Spartan-VI page on the Dynamic Engineering website. If you want some of the new features, and have already purchased hardware, we will support you with a FLASH update.

Vendorld = xDCBACardId = x0077.



### Programming

PCIe-Spartan-VI is tested in a Windows environment. We use the Dynamic Engineering Driver to do the low level accesses to the hardware. We use MS Visual Studio in conjunction with the driver to write our test software. The driver package is included with your purchase.

The drivers take care of discovery, and the UserAp allows the client to select which installed board is selected for use.

If you are writing your own driver it is suggested to get the engineering kit and the Linux version of the SW. Usually, the defines and perhaps some of the code can be reused in your effort.

The following pages have the memory maps and bit maps for the Bus Controller "fixed" features and User FPGA reference design features. The User FPGA memory map is subject to change as you implement your design.



### **Address Map**

#### **Controller Base Address Map**

PcieSpartanVI_BASE_BASE	0x0000 // 0 Base control register
PcieSpartanVI_BASE_USER_SWITCH	0x0004 // 1 User switch read port DIP switch read
PcieSpartanVI_BASE_XILINX_REV	0x0004 // 1 Xilinx revision read port
PcieSpartanVI_BASE_XILINX_DSN	0x0004 // 1 Xilinx Design Number read port
PcieSpartanVI_BASE_STATUS	0x0008 // 2 status Register offset
PcieSpartanVI_BASE_JTAG	0x000C // 3 JTAG interface Port
PcieSpartanVI _CH0	0x0050 // 20 starting address for channel 0
PcieSpartanVI _CH1	0x00A0 // 40 starting address for channel 1
PcieSpartanVI _CH2	0x00F0 // 60 starting address for channel 2
PcieSpartanVI _CH3	0x0140 // 80 starting address for channel 3
PcieSpartanVI _CH4	0x0190 // 100 starting address for channel 4
PcieSpartanVI _CH5	0x01E0 // 120 starting address for channel 5
PcieSpartanVI _CH6	0x0230 // 140 starting address for channel 6
PcieSpartanVI _CH7	0x0280 // 160 starting address for channel 7

Figure 3

PCIe-Spartan-VI Controller Base Address Map

The address map provided is for the local decoding performed within PCIe-Spartan-VI Bus Controller. The addresses are all offsets from a base address. The base address and interrupt level is provided by the host in which the PCIe-Spartan-VI is installed.

#### **Controller Port Address Map**

0x00000000 // 0 General control register
0x00000004 // 1 Interrupt status port
0x00000004 // 1 Interrupt clear port
0x00000008 // 2 Write DMA dpr physical PCI address
0x00000008 // 2 Tx FIFO count read port
0x0000000C // 3 Read DMA dpr physical PCI address
0x0000000C // 3 Rx FIFO count read port
0x00000010 // 4 FIFO offset for single word access R/W
0x00000014 // 5 Tx almost empty count register - used
for Urgent and pulsed interrupt
0x00000018 // 6 Rx almost full count register
0x00000028 // 10 Tx almost empty level register - used
for level based FIFO interrupt
0x00000040 // 16 Rx almost full level register

Figure 4

PCIe-Spartan-VI Controller Port Address Map

There are 8 channels each with the memory map shown above. The offset to each channel relative to the base address as shown in the Base Address Map. Decodes 0-19 are reserved for the Base, 20-39 to channel 0, 40-59 to channel 1 etc.



The host system will enumerate to find the assets installed during power-on initialization. Interrupts are requested by the configuration space. Third party utilities can be useful to see how your system is configured. The interrupt level expected and style is also set in the registry. Dynamic Engineering recommends using the Dynamic Engineering Driver to take care of initialization and device registration.

Once the initialization process has occurred, and the system has assigned an address range to PCIe-Spartan-VI, software will need to determine what the address space is. We refer to this address as base in our software.

The next step is to initialize PCIe-Spartan-VI. The local Controller registers need to be configured as well as the registers within the User FPGA.

The following address maps for the User FPGA are based on the reference design. Your design implementation may change the addresses and bitmaps.

0x8000 // PcieSpartanVI Base Set Addr(15) to use 32K dedicated to User control bus
0x0000 // 0 User base address for PLL access
0x0004 // 1 User address for LED access Bit 4 enables
to SW control
0x0008 // 2 User address for Status Register in Base
0x0010 // 4 User address for TTL Data output and read-
back of IO level 11-0
0x0014 // 5 User address for TTL Data Enables 11-0
0x001C // 7 PcieSpartanVI Base Write Master Count,
Read selected count
0x0028 // 10 User address for PLL FIFO
0x0050 // 20 User address pointer offset for channel 0
0x00A0 // 40 User address pointer offset for channel 1
0x00F0 // 60 User address pointer offset for channel 2
0x0140 // 80 User address pointer offset for channel 3
0x0190 // 100 User address pointer offset for channel 4
0x01E0 // 120 User address pointer offset for channel 5
0x0230 // 140 User address pointer offset for channel 6
0x0280 // 160 User address pointer offset for channel 7

#### **User Base Address Map**

Figure 5

PCIe-Spartan-VI User Base Address Map



#### **User Port Address Map**

PcieSpartanVI_USER _CH_BASE	0x0000 // 0 User address for Base Register
PcieSpartanVI_USER _CH_STATUS	0x0004 // 1 User address for Status
PcieSpartanVI_USER _CH_SP	0x0008 // 2 User address for Spare Register
PcieSpartanVI_USER _UB_FIFO_CNT	0x000C // 3 User address for User to Cont FIFO Count
PcieSpartanVI_USER _BU_FIFO_CNT	0x0010 // 4 User address for Cont to User FIFO Count
PcieSpartanVI_USER _FIFO_WR	0x0014 // 6 User address for Writing RX FIFO
PcieSpartanVI_USER _FIFO_RD	0x0018 // 6 User address for Reading TX FIFO

Figure 6

PCIe-Spartan-VI User Port Address Map

Dynamic Drivers provide a GPB write and GPB read utility which allows the user to pass an offset and data to write or an offset to read and return a LW. The utility assumes the offset to the User FPGA. User SW can use the above definitions or self-defined mapping to build a pointer to the intended offset. For example:

PcieSpartanVI\_USER\_CH\_7 + PcieSpartanVI\_USER\_CH\_STATUS is the offset for the Status register in channel 7 of the User FPGA.



### **Register Definitions**

#### **Bus Controller Base Register Definitions**

#### PcieSpartanVI\_BASE\_BASE

[0x0000 Main Control Register Port read/write]

DATA BIT	BASE REGISTER DESCRIPTION	
0	BASE_USER_INT_MASK	
15	BASE_USER_RST	

Figure 7

PCIe-Spartan-VI Xilinx Base Control Register

BASE\_USER\_INT\_MASK : Set to enable interrupt from User FPGA design. Default is disabled. When '1' the master enable is "enabled". Further enables within the User FPGA may need to be programmed.

BASE\_USER\_RST: Set to reset User device. The decode for the address is anded with the data bit to create a pulse on the Control Reset signal. This is the reset for the GPB bus and frequently used as the system reset for the User FPGA design.



#### PcieSpartanVI\_BASE\_ID

[0x04 Switch and Design number port	read only	
DATA BIT	DESCRIPTION	
31-24	Design Revision Minor	
23-16	Design Number Major	
15-8	PCI Core Revision	
7-0	DIP switch	

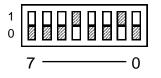
[0x04 Switch and Design number port read only]

Figure 8

PCIe-Spartan-VI Xilinx ID Register

The ID register is made up of the DIP switch port plus design revision, and PCI Core Revision.

The DIP Switch is labeled for bit number and '1' '0' in the silk screen. The DIP Switch can be read from this port and used to determine which PCIe-Spartan-VI is which in a system with multiple cards installed. The DIP switch can also be used for other purposes – software revision etc. The switch shown would read back 0x12.



The Design Revision Major.Minor allow for tracking the released FPGA versions. The Major revision is updated for large changes to the design – for example changing an EOL part with SW implications. The Minor revision is used during development of the Major change plus tracking small enhancements.

The PCI revision is the revision of the core in use. It is unlikely to change.

Revision Major.Minor 1.1 5/23/24 Initial Release



#### PcieSpartanVI\_BASE\_STATUS

[0x0008 status register read only]

	STATUS REGISTER	
DATA BIT	DESCRIPTION	
24.44	Crock	
31-14	Spare	
13	UB Ref Lock	
12	BU Ref Lock	
11-10	Spare	
9	User Int Masked	
8	User Int	
7	Ch7 Int Status	
6	Ch6 Int Status	
5	Ch5 Int Status	
4	Ch4 Int Status	
3	Ch3 Int Status	
2	Ch2 Int Status	
1	Ch1 Int Status	
0	Ch0 Int Status	

#### Figure 9

PCIe-Spartan-VI Status Port

The masks for the Controller Port interrupts are in the Port. When Ch(0:7) Int Status are set an interrupt is pending from the channel or channels indicated.

The User FPGA can generate an interrupt request. A second mask is provided within the Controller for this Interrupt request to allow for polling. See the Controller Base Register. User Int is the unmasked signal. User Int Masked is the version to check for an active interrupt, and is after the mask.

The two Lock Bits indicate the local clock distribution is in "Lock" meaning the signal has been acquired properly. UB is the reference clock from User FPGA to Bus FPGA and the BU version from the Bus FPGA to the User FPGA. Both need to be locked to operate properly. The BU lock is completely self-contained. The UB Lock could be out of lock if the User Design does not implement the clock properly.



#### PcieSpartanVI\_BASE\_JTAG

[0x000C program the User FPGA]

DATA BIT	DESCRIPTION	
31	MUX select	
30-29	spare	
28	program	
27-6	spare	
5	Done Read Only	
4	Init Read Only	
3	TDO	
2	TMS	
1	TCK	
0	TDI	

Figure 10

PCIe-Spartan-VI Programming port

TDI is set or cleared to create the data pattern to load JTAG data into the TDI pin on the User device. TCK and TMS are used in conjunction with TDI to create the programming stream. If programming the QSPI the FPGA will need to be programmed with the JTAG to QSPI module – automatic from the Impact program.

TDO is synchronized to the local clock and available for read-back.

The MUX Select when '0' provides a path from the Header to the User device for direct programming from the platform programming adapter. When set to '1' the mux provides a path from this device to the User FPGA allowing the User device to be programmed directly or the data stored into QSPI to load at power up.

Done and Init can be polled during initialization of the User FPGA to see the proper sequence is happening or just the final state is correct. Done and Init are driven low by the User FPGA, Init is released, then Done is released once programming is completed.

When the Program control bit is set to '1' the User FPGA Program control input is driven low '0'. When cleared the Program bit is released and an external pull-up will return high. The Program bit is automatic – the data bit is ANDed with the Select for the register to create a pulse. No need to clear. Please note: when the QSPI is programmed the sequence includes reloading the FPGA from the programmed QSPI.

The reference design initializes with x5 on the User LEDs. The LED test changes to xA. The PLL test toggles and returns to the initial value. A user design can change the value to provide a visual that the new program is successfully implemented.



#### **Bus Controller Port Register Definitions**

#### PcieSpartanVI\_CH\_CNTRL

[0x0000 read/write]

	Channel Control Register
DESCRIPTION	Bit
CNTRL_TX_FIFO_RST	0x00000001 //0 set to clear FIFO's
CNTRL_RX_FIFO_RST	0x00000002 //1 set to clear FIFO's
CNTRL_FF_TEST	0x00000004 //2 bypass mode
CNTRL_MINTEN	0x0000008 //3 channel interrupt enable
CNTRL FORCE INT	0x00000010 //4 channel based force interrupt function
CNTRL_DMA_WREN	0x00000020 //5 DMA interrupt enable burst in
CNTRL DMA RDEN	0x00000040 //6 DMA interrupt enable burst out
CNTRL_PAC4_TXSTART	0x00010000 //16 enable Tx State machine
CNTRL_PAC4_RXSTART	0x00020000 //17 enable Rx State Machine
CNTRL_PAC4_TX_FF_AMT_INT	0x00040000 //18 enable Transmit FIFO Almost Empty
	Interrupt based on Pulse
CNTRL_PAC4_RX_FF_AFL_INT	0x00080000 //19 enable Receive FIFO Almost Full
	Interrupt based on Pulse
CNTRL_PAC4_OFL_INT	0x00100000 //20 enable Receive OverFlow interrupt
CNTRL_PAC4_TX_FF_AMT_INT_LVL	0x00200000 //21 enable Transmit FIFO Almost Empty
	Interrupt based on Level
CNTRL_PAC4_RX_FF_AFL_INT_LVL	0x00400000 //22 enable Receive FIFO Almost Full
	Interrupt based on Level
Figure 11	PCIe-Spartan-VI Controller Channel Control Register

CNTRL\_TX\_FIFO\_RST and CNTRL\_RX\_FIFO\_RST when set '1' cause the respective memories to be reset to the empty state. Clear to '0' for normal operation.

CNTRL\_FF\_TEST when set '1' routes data from the TX FIFO to the RX FIFO for loopback testing within the Controller. Can be used with DMA or target accesses. Examples of how to operate are in the reference software.

CNTRL\_MINTEN is the master interrupt enable. This bit needs to be set to have the particular channel able to cause an interrupt. In addition, individual masks for the interrupts are provided.

CNTRL\_FORCE\_INT when set causes an interrupt from the channel. Useful for debugging and checking interrupt handling software.

CNTRL\_DMA\_WREN [Burst In] and CNTRL\_DMA\_RDEN [Burst Out] when set enable DMA interrupts to be active. BurstIn refers to data coming from system memory to



PCIe-Spartan-VI and BurstOut refers to data going to system memory.

CNTRL\_PAC4\_TX\_FF\_AMT\_INT and CNTRL\_PAC4\_RX\_FF\_AFL\_INT when set enable interrupts based on the FIFO level transition. From not Almost Full to Almost Full for the RX FIFO, and/or not Almost Empty to Almost Empty for the TX FIFO. These interrupts can be handy if using small DMA transfers that need to be managed. The interrupts are cleared by writing to the associated status register "sticky" bits.

CNTRL\_PAC4\_OFL\_INT enables the interrupt on receive FIFO overflow. This condition is in theory "impossible" since there is HW flow control.

CNTRL\_PAC4\_TX\_FF\_AMT\_INT\_LVL and CNTRL\_PAC4\_RX\_FF\_AFL\_INT\_LVL when set enable interrupts based on the FIFO level. When Almost Full for the RX FIFO and/or Almost Empty for the TX FIFO. These interrupts can be handy if using small DMA transfers that need to be managed. These interrupts stay set until the level is returned above or below the threshold. The masks can be used to disable.

CNTRL\_PAC4\_TXSTART and CNTRL\_PAC4\_RXSTART enable the data transfer state-machines for the transmit [to User] and receive [from User] byte lanes. Normally both are enabled. For some test situations one or both can be disabled.

No interrupts are associated with the state-machines directly. Interrupts are available for the completion of a DMA or based on programmable FIFO levels to allow for user control.



#### PcieSpartanVI\_CH\_STATUS

[0x004 read/write]

DESCRIPTION	Channel Control Register Bit
STAT_TX_FIFO_MT	0x00000001 //0 set when TX FIFO is empty
STAT_TX_FIFO_AE STAT_TX_FIFO_FULL	0x00000002 //1 set when TX FIFO is Almost Empty 0x00000004 //2 set when TX FIFO is Full
STAT_RX_FIFO_MT	0x00000010 //4 set when RX FIFO is Empty
STAT_RX_FIFO_AF STAT_RX_FIFO_FULL	0x00000020 //5 set when RX FIFO is Almost Full 0x00000040 //6 set when RX FIFO is Full
STAT_TX_AMT_INT	0x00000100 //8 Transmit Almost Empty Int Occurred
STAT_RX_AFL_INT STAT_TX_FF_INT_LAT	0x00000200 //9 Receive Almost Full Int Occurred 0x00000400 //10 Transmit Almost Full Int Occurred
STAT_RX_FF_INT_LAT	0x00000800 //11 Receive Almost Full Int Occurred
STAT_BURSTIN_ERR	0x00001000 //12 write DMA error
STAT_BURSTOUT_ERR STAT_WR_DMA_INT	0x00002000 //13 read DMA error 0x00004000 //14 write DMA Interrupt
STAT_RD_DMA_INT	0x00008000 //15 read DMA Interrupt
STAT_RX_OVFL_LAT	0x00080000 //19 Rx OverFlow Latched
STAT_BO_IDLE	0x00400000 //22 Burst Out Idle
STAT_BI_IDLE	0x00800000 //23 Burst In Idle
LOC_INT	0x40000000 //30 channel interrupt
STAT_INT_ACTIVE	0x80000000 //31 channel interrupt is active
Figure 12	PCIe-Spartan-VI Controller Channel Status Register

Figure 12

PCIe-Spartan-VI Controller Channel Status Register

<u>Transmit FIFO Empty</u>: When a one is read, the transmit data FIFO contains no data; when a zero is read, there is at least one data word in the FIFO.

<u>Transmit FIFO Almost Empty</u>: When a one is read, the number of data words in the transmit data FIFO is less than or equal to the value written to the TX\_AMT\_LVL register; when a zero is read, the FIFO level is more than that value.

<u>Transmit FIFO Full</u>: When a one is read, the transmit data FIFO is full; when a zero is read, there is room for at least one more data word in the FIFO.

Please note with the Receive side status; the status reflects the state of the FIFO and does not take the 4 deep pipeline into account. For example, the FIFO may be empty and there may be valid data within the pipeline. The data count is the combined FIFO and pipeline value and can be used for read size control.



<u>Receive FIFO Empty</u>: When a one is read, the receive data FIFO contains no data; when a zero is read, there is at least one data word in the FIFO.

<u>Receive FIFO Almost Full</u>: When a one is read, the number of data words in the receive data FIFO is greater or equal to the value written to the RX\_AFL\_LVL register; when a zero is read, the FIFO level is less than that value.

<u>Receive FIFO Full</u>: When a one is read, the receive data FIFO is full; when a zero is read, there is room for at least one more data-word in the FIFO.

STAT\_TX\_AMT\_INT: When a one is read, the transmit FIFO almost empty is asserted. This is a level based interrupt. The status is before the mask to allow for non-interrupt driven SW operation. When the FIFO level is below the programmed level this bit is asserted.

STAT\_TX\_FF\_INT\_LAT: When a one is read, the transmit FIFO almost empty has been asserted. This is a triggered interrupt. The status is before the mask to allow for non-interrupt driven SW operation. When the FIFO level has dropped below the programmed level the status is captured and held. Clear by writing back to this bit.

STAT\_RX\_AFL\_INT: When a one is read, the receive FIFO almost full is asserted. This is a level based interrupt. The status is before the mask to allow for non-interrupt driven SW operation. When the FIFO level is above the programmed level this bit is asserted.

STAT\_RX\_FF\_INT\_LAT: When a one is read, the receive FIFO almost full has been asserted. This is a triggered interrupt. The status is before the mask to allow for non-interrupt driven SW operation. When the FIFO level has risen the programmed level the status is captured and held. Clear by writing back to this bit.

<u>Write/Read DMA Error Occurred</u>: [STAT\_BI\_ERR, STAT\_BO\_ERR] When a one is read, a write or read DMA error has been detected. This will occur if there is a target or master abort or if the direction bit in the next pointer of one of the chaining descriptors is incorrect. A zero indicates that no write or read DMA error has occurred. These bits are latched and can be cleared by writing back to the Status register with a one in the appropriate bit position.

<u>BO and BI Idle</u> [STAT\_BO\_IDLE, STAT\_BI\_IDLE] are Burst Out and Burst In IDLE state status for the Receive and Transmit DMA actions. The bits will be 1 when in the IDLE state and 0 when processing a DMA. A new DMA should not be launched until the State machine is back in the IDLE state. Please note that the direction implied in the name has to do with the DMA direction – Burst data into the card for TX and burst data out of the card for Receive.



<u>Write/Read DMA Interrupt Occurred</u>: [STAT\_WR\_DMA\_INT, STAT\_RD\_DMA\_INT] When a one is read, a write/read DMA interrupt is latched. This indicates that the scatter-gather list for the current write or read DMA has completed, but the associated interrupt has yet to be processed. A zero indicates that no write or read DMA interrupt is pending.

<u>Channel Interrupt Active</u>: When a one is read, it indicates that a system interrupt is potentially asserted caused by an enabled channel interrupt condition. A zero indicates that no system interrupt is pending from an enabled channel interrupt condition. The Board level master interrupt enable will also need to be asserted to allow the active channel interrupt to become an interrupt request.

<u>Local Interrupt</u>: When a one is read, it indicates that any of the masked conditions other than DMA are active and enabled. Local Interrupt is or'd with the DMA interrupt sources to create the Channel Interrupt Active signal and to request the Interrupt.

STAT\_RX\_OVFL\_LAT is set if the RX FIFO is overwritten. Should never occur as there is flow control in place. However, since this is a user design it is possible for user changes to break the flow control. Clear by writing back with this bit set.



#### PcieSpartanVI\_CH\_WR\_DMA\_PNTR

[0x008 Write only]

DMA	A Pointer Address Register
Data Bit	Description
31-2	First Chaining Descriptor Physical Address
1	direction [0]
0	end of chain

Figure 13

PCIe-Spartan-VI Controller Channel Write DMA Register

This write-only port is used to initiate a scatter-gather write [TX] DMA. When the address of the first chaining descriptor is written to this port, the DMA engine reads three successive long words beginning at that address. Essentially this data acts like a chaining descriptor value pointing to the next value in the chain.

The first is the address of the first memory block of the DMA buffer containing the data to read into the device, the second is the length in bytes of that block, and the third is the address of the next chaining descriptor in the list of buffer memory blocks. This process is continued until the end-of-chain bit in one of the next pointer values read indicates that it is the last chaining descriptor in the list.

All three values are on LW boundaries and are LW in size. Addresses for successive parameters are incremented. The addresses are physical addresses the HW will use on the PCI bus to access the Host memory for the next descriptor or to read the data to be transmitted. In most OS you will need to convert from virtual to physical. The length parameter is a number of bytes, and must be on a LW divisible number of bytes.

Status for the DMA activity can be found in the channel control register and channel status register.

Notes:

- 1. Writing a zero to this port will abort a write DMA in progress.
- 2. End of chain should not be set for the address written to the DMA Pointer Address Register. End of chain should be set when the descriptor follows the last length parameter.
- 3. The Direction should be set to '0' for Burst In DMA in all chaining descriptor locations.



#### PcieSpartanVI\_CH\_RD\_DMA\_PNTR

[0x00C Write only]

DMA	Pointer Address Register
Data Bit	Description
31-2	First Chaining Descriptor Physical Address
1	direction [1]
0	end of chain

Figure 14

PCIe-Spartan-VI Controller Channel Read DMA Register

This write-only port is used to initiate a scatter-gather read [RX] DMA. When the address of the first chaining descriptor is written to this port, the DMA engine reads three successive long words beginning at that address. Essentially this data acts like a chaining descriptor value pointing to the next value in the chain.

The first is the address of the first memory block of the DMA buffer to write data from the device to, the second is the length in bytes of that block, and the third is the address of the next chaining descriptor in the list of buffer memory blocks. This process is continued until the end-of-chain bit in one of the next pointer values read indicates that it is the last chaining descriptor in the list.

All three values are on LW boundaries and are LW in size. Addresses for successive parameters are incremented. The addresses are physical addresses the HW will use on the PCI bus to access the Host memory for the next descriptor or to read the data to be transmitted. In most OS you will need to convert from virtual to physical. The length parameter is a number of bytes, and must be on a LW divisible number of bytes.

Status for the DMA activity can be found in the channel control register and channel status register.

Notes:

- 1. Writing a zero to this port will abort a write DMA in progress.
- 2. End of chain should not be set for the address written to the DMA Pointer Address Register. End of chain should be set when the descriptor follows the last length parameter.
- 3. The Direction should be set to '1' for Burst Out DMA in all chaining descriptor locations.



#### PcieSpartanVI\_CH\_TX\_FIFO\_CNT

[0x008 Port Read only]

	TX FIFO Data Count Port	
Data Bit 31-0	<b>Description</b> TX Data Words Stored	

Figure 15

PCIe-Spartan-VI Controller Channel TX FIFO Count

This read-only register port reports the number of 32-bit data words in the transmit FIFO. The TX FIFO has 8K-1 32 bit locations.

#### PcieSpartanVI\_CH\_RX\_FIFO\_CNT

[0x00C Port Read only]

RX	FIFO Data Count Port	
Data Bit 31-0	<b>Description</b> RX Data Words Stored	
010	NX Data Werds Otored	

Figure 16

PCIe-Spartan-VI Controller Channel RX FIFO Count

This read-only register port reports the number of 32-bit equivalent data words in the receive pipeline. There are three components – the initial 16Kx8 FIFO, the second 16Kx8 - 4Kx32 FIFO and the DMA pipeline. The count from the x8 initial FIFO is divided by 4 and synchronized prior to adding to the secondary FIFO and pipeline counts. Due to the conversion, there may be some jitter in the lower bits of this count when read during active transfer. When static the count will be correct. The total for the pipeline is 8K+2.

Note: if non-32 bit lengths are transferred between the User and Bus devices the remainder will not show in the count [right shift 2 places for divide by 4] and remain in the pipeline until more data is presented.



#### PcieSpartanVI\_CH\_FIFO

[0x010 Port Read/Write]

	<b>RX and TX FIFO Port</b>	
Data Bit 31-0	Description FIFO data word	

Figure 17

PCIe-Spartan-VI Controller Channel FIFO Access

This port is used to make single-word accesses into the TX and out of the RX FIFO. Please note that reading is from the RX FIFO and writing is to the TX FIFO. Unless Bypass mode is established the data will not match.

#### PcieSpartanVI\_CH\_TX\_AMT

[0x014 Port Read/Write]

	TX Almost-	Empty Pulse Register
<b>Data</b> 31		Description TX FIFO Almost-Empty Level

Figure 18

PCIe-Spartan-VI Controller Channel TX Almost Empty Pulse

This read/write port accesses the transmitter almost-empty level register. When the number of data words in the transmit data FIFO is less than this value, the almost-empty status bit will be set. Used for the Pulsed interrupt and shown as a status bit.



#### PcieSpartanVI\_CH\_RX\_AFL

[0x018 Port Read/Write]

RX	Almost-Full Pulse Register
Data Bit	Description
31-0	RX FIFO Almost-Full Level

Figure 19

PCIe-Spartan-VI Controller Channel RX Almost Full Pulse

This read/write port accesses the receiver almost-full level register. When the number of data words in the receive data FIFO is greater than this value, the almost-full status bit will be set. The mask is valid for a size matching the depth of the FIFO. This value is applied to the entire FIFO chain [8K+2]. Used for the Pulsed interrupt and status.

#### PcieSpartanVI\_CH\_TX\_AMT\_LVL

[0x028 Port Read/Write]

тх	Almost-Empty Level Register	
Data Bit 31-0	Description TX FIFO Almost-Empty Level	

Figure 20

PCIe-Spartan-VI Controller Channel TX Almost Empty Level

This read/write port accesses the transmitter almost-empty level register. When the number of data words in the transmit data FIFO is less than this value, the almost-empty status bit will be set. Used for the Level based interrupt and status.



#### PcieSpartanVI\_CH\_RX\_AFL\_LVL

[0x040 Port Read/Write]

RX	Almost-Full Level Register	
Data Bit 31-0	Description RX FIFO Almost-Full Level	

Figure 21

PCIe-Spartan-VI Controller Channel RX Almost Full Level

This read/write port accesses the receiver almost-full level register. When the number of data words in the receive data FIFO is greater than this value, the almost-full status bit will be set. This value is applied to the entire FIFO chain [8K+2]. Used for the Level based interrupt and status.



#### User FPGA Base Address Map

The User FPGA is completely programmable – the address map above and definitions below only have meaning if the reference VHDL is used as a starting point for your design. The reference software and reference User hardware implementation are used to perform the ATP on each board prior to shipment. The Engineering kit can include the HDEterm100 and a cable to interconnect the PCIe-Spartan-VI with the HDEterm100. For more information, please refer to the web page.

PcieSpartanVI\_BASE\_USER\_ACCESS defines the offset from the Bus FPGA base address to the User FPGA Base address. Dynamic Drivers automatically include this offset. If you are doing your own driver you will need to include this offset. All other offsets shown are relative to the User base address.

#### PcieSpartanVI\_USER\_BASE

[0x000 Main Control Register Port read/write]

SCRIPTION tNum rceInt
rceInt
John
ntEn
are
PII
are
L Programming Enable
unter Select
L_CLK_EN_(70)

Figure 22

PCIe-Spartan-VI User Base Control Register

PCIe-Spartan-VI has 8 PLL devices which are programmed to produce the desired frequency with an i2c bus. Each PLL has a common data pin and independent clocks. The PLLs also have independent references.

PLL\_CLK\_EN\_(7..0).when set selects that PLL for loading or reading. Reading is required to only have one PLL selected. Writing can be done in parallel or with separate writes to each device. For parallel writes to work the address of each PLL must be the same.

Counter Select is used to pick the counter to read-back with the "USER CNT" register. Once the PLLs are programmed the outputs can be used to count and check if the expected frequency is received.



The reference design contains the logic and software required to program the PLLs and to read the programmed frequency back. The software to determine the frequency command words is available from Cypress Semiconductor. The part number is CY22393FXI. Cypress has a utility available for calculating the frequency control words for the PLLs. <u>https://www.dyneng.com/\_Download/Utilities/CyberClocks.zip</u> is the URL for the Cypress software used to calculate the PLL programming words.

The PLLs respond to one of two addresses [only one works]. As part of the ATP the reference software determines the address of each PLL and stores into an array. The remaining functions within the PLL test section use the data from the array to access each of the PLLs. Functions for loading, reading, comparing the read-back and expected load pattern, plus counting and checking against the expected count are provided. The software is part of the engineering kit and can be ported to your application.

PLL Programming Enable when set ('1') enables the i2c state-machine to begin operation. The state-machine will read the data from the FIFO and transfer to or read from the selected PLL as directed by the initial instruction read from the FIFO. Please refer to the "USER PLL FIFO" section for the header information.

CIrPII when set '1' resets the PLL interface – FIFOs etc. When '0' the interface is in standard mode. Can set and clear on consecutive GPB accesses.

MIntEn is the master interrupt enable for User FPGA. Default is disabled. When '1' the master enable is "enabled".

ForceInt when '1' and the MintEn is enabled causes an interrupt to be generated. This bit is useful for software debugging.

Please note the Controller enable for the User Interrupt must also be enabled.

InstNum is a R/W field where the driver can store the instance of the board. Useful for multiple board implementations.



### PcieSpartanVI\_USER\_LED

[0x004 read/write]

LED REGISTER		
DATA BIT	DESCRIPTION	
31-5 4 3 - 0	spare LED Enable LED Control	

Figure 23

PCIe-Spartan-VI User LED Control Register

LED Control when set ('1') and LED Enable is set ('1') causes the individual LEDs to be illuminated. When '0' the LED is off. When LED Enable is '0' the LED pattern is controlled by the HW strapped default. The FLASH based reference design is set to "0x5". Please note that the HW straps are counter-intuitive as the LED is on when '0' is supplied and off when '1' is supplied. The SW path has the inversion built in.

### PcieSpartanVI\_USER\_STATUS

[0x008 Status Register read/write]

Status Register	
DATA BIT	DESCRIPTION
31-24	DesignId
23-16	Int(7-0)
15-11	spare
10-8	PIIPckDnCnt
7	PIINakLat
6	PIIPacketDoneLat
5	PIIReadFifoMt
4	PIIWriteFifoMt
3	PIIIdle
2	PIIEn
1	LocalInt
0	spare

Figure 24

PCIe-Spartan-VI User Status Register

Int(7-0) when set indicate that a channel interrupt is pending. If the master interrupt is enabled the interrupt request from the User flows through to the Controller where it can be passed onto the system. If the master enable is not set the INTx status can be used for polling. Additional masking is provided within the channels. INTO is associated with



Channel 0 etc. LocalInt is set if any of the INTx bits are set [before the master interrupt enable].

DesignId is set to allow the user software to read the Design Number and determine how to interact with the design.

When PIIEn is set the PLLCounting test is in operation. The test is started by writing the control count to the master counter. When the master counter hits 0x00 the test completes. PIIEn is set while the counter is operating.

PIIIdle is set when the PLL controller is in the IDLE state. This bit can be polled to track status on the loading of the PLLs by the state-machine. Please note the state-machine will pass through the IDLE state during multiple packet operations.

PllWriteFifoMt is '1' when the FIFO associated with moving data to the PLL is empty.

PllReadFifoMt is '1' when the FIFO associated with data read from the PLL is empty. Please note: this FIFO will receive data when testing addresses etc.

PIIPacketDoneLat is set and held when the PLL state-machine is done processing a packet. The bit is held until the same bit is written to – that is write to the status port with this bit '1' to clear.

PIINakLat is set when the PLL does not respond to an access attempt. This is normal during discovery and abnormal once the addresses are known. Clear by write-back to the status port with this bit set.

PIIPckDnCnt is a 3 bit field which is incremented when a packet completes. During PLL programming address offsets are used which makes one load take more than one packet. The count is useful for determining when the complete PLL programming cycle has happened. Clear to 0x00 before using to check a load. Clear by CIrPII or writing back to the status register and clearing the PIIPacketDoneLat [also clears this field].



### PcieSpartanVI\_USER\_TTL\_DAT

[0x010 TTL Data Port read/write]

TTL Data Register		
DATA BIT	DESCRIPTION	
11-0	TTL Data 11-0	

Figure 25

PCIe-Spartan-VI TTL Data Register

The TTL data pattern written to this port will be loaded to the output side for the bits enabled in the TTL\_EN register. All bits when read are from the IO input. The IO may or may not match the output definition depending on which bits are enabled and what is connected on the IO side of the interface.

#### PcieSpartanVI\_USER\_TTL\_EN

[0x014 TTL Data Enable Port read/write]

485/LVDS IO REGISTERS		
DATA BIT	DESCRIPTION	
11-0	TTL Data Enable	

Figure 26

PCIe-Spartan-VI TTL Data Enable Register

The TTL IO each have individual enables. The bits set in this register are enabled to drive onto the cable side. Please note that the enables are inverted in HW to provide an active low enable at the '125 buffer devices. SW sets/reads as '1' = enabled, '0' = disabled [read this bit from the cable]. If the designer enables the bits in this register the IO are driven with 24 mA [drive / sink] buffers. If the bits are tied to the Output register the IO will be enabled for active low and disabled for active high signals – open drain operation. The reference for open drain operation is set by the header. The parts are also sourced by the selected voltage.



## Application Note: Spare IO

Frequently a system will need some dedicated IO but not all of the IO. The reference design has all of the differential IO defined in the Channels, and all of the TTL IO defined in the Base. The mix can easily be changed to put TTL in the channels or Differential in the channels. "extra" IO can be used as a parallel port or other use. If not used please tie off to be controlled as an input to reduce system noise and avoid potential conflicts on the cabling.

#### PcieSpartanVI\_USER\_CNT

[0x01C Count read-back port]

Counter Port		
DATA BIT	DESCRIPTION	
15-0	Count	

Figure 27

PCIe-Spartan-VI Counter Results

There are 25 counters implemented to check the PLL programming. The Counter results to read are selected in the User Base register and read from this register. The 0x00 selection is the control count which will be 0 at the test completion. 0x01 corresponds to [silk screen] PLL1 output A. 0x02 to PLL1 output B etc, 0x18 = PLL8 output C.

The reference software has two counting tests. One test programs the PLLs to convert the 66.666 reference to 10 MHz on all channels and checks that all are within tolerance. The second test programs the PLLs to all different frequencies 10-33 MHz and again checks for the expected frequency to be measured via count. Most designs won't need this logic. It can be removed to create space in the FPGA if you are using the majority of the FPGA assets.

Design note: all 24 clocks are on FPGA clock inputs. Please see the PLL pin definitions in the pinout table. You may want to do some floor planning to have the clocks in the same vicinity as the logic. If only using a few clocks this is not necessary.



### PcieSpartanVI\_USER\_PLL\_FIFO

[0x028 PLL Data R/W port]

TTL IO REGISTERS		
DATA BIT	DESCRIPTION	
31-0	Data to/from PLL	

Figure 28

PCIe-Spartan-VI PLL FIFO

The transmit FIFO is monitored by the PLL state-machine. When the FIFO is written to the first word is read by the state-machine and parsed. The first word contains the mode on bit 0, address on 7-1, length on 15-8 [1-255], and the first byte or two to transfer. If multiple bytes – 3 or more are to be transferred the SW will need to make sure the data is in the FIFO for the 2<sup>nd</sup> LW before the end of the processing of the 2<sup>nd</sup> byte or an underflow condition will be detected. If your system timing is tough to manage it is suggested to disable the SM, load the FIFO and then enable the SM. A status bit for the idle condition is available to allow SW to know when the SM has responded to the disable.

The Length is the number of bytes in the data portion of the message + 1.

Please note: The PLLs have two data sets written to two address offsets per PLL programmed. The UserAp automatically converts the .jed file from the Cypress tool and generates the local buffers with the hex data to load to the PLL. The application software loads the FIFO with the correct address, length and data x2 for a complete programming operation.

The State-machine will parse the message and write or read based on bit 0. In either case the address and R/W are transmitted. An ACK is looked for from the Target. If a Write the data is then transmitted with the ACK being checked after each byte. Clocking is continuous until the message is completed. If a read is implemented, clocks are generated without data after the address. Data is captured during the high portion of the clock cycle, and the Master asserts the ACK until the last byte where a NAK is asserted. Data is stored into the receive FIFO in this case.

The reference software has examples of working with the PLLs and controlling HW.



## **User Channel Address Map**

### PcieSpartanVI\_USER\_CH\_BASE

[0x000 Main Control Register Port read/write]

Channel Base Register	
DATA BIT	DESCRIPTION
31-28	spare
27	RxloEn
26	TxloEn
25	spare
24	RxFfAFIIntEnLvl
23	TxFfAMtIntEnLvI
22	TxOvFIIntEn
21	TxFfIntEn
20	RxFfIntEn
19	BUStart
18	UBStart
17	TxStart
16	RxStart
15-5	spare
4	IntForce
3	MIntEn
2	Bypass
1	RxRst
0	TxRst

Figure 29

PCIe-Spartan-VI Channel Base Register

TxRst When set ('1') causes Tx HW to be reset. FIFOs, state-machines etc.

RxRst when set ('1') causes Rx HW to be reset including FIFOs etc.

Bypass when set ('1') causes the HW to transfer data from the input data lane FIFO to the output data lane FIFO. The reference software has a loop-back test checking this data path. It is a good indicator that the User FPGA is properly loaded and the data lanes are functional. It is recommended to keep this logic in place to allow for BIT.

MIntEn when set ('1') allows the enabled channel status to cause interrupts via the connection to the Bus Controller. The Master in the Bus Controller also needs to be set for this to work.

IntForce when set '1' and MintEn is set, causes an interrupt to be requested. Mainly used as a development tool to check interrupt paths and to simulate conditions when the external system is not in place.



TxStart when set enables data transfer to IO. Data in Tx FIFO sent to the IO under state-machine control.

RxStart when set enables data transfer from IO into receiver memory. Receiving from external system.

UBStart enables the transfer of data from the User FIFO to the Bus FIFO for the port. If both the RxStart and UBStart are enabled data is moved from IO to the Bus FPGA. With DMA enabled the transfer can be direct into system memory.

BUStart is a place holder and not currently implemented. Current reference design is always enabled to receive data into the holding FIFO within the User design. If the TxStart is not enabled data will fill the holding FIFO and stop transfer due to the flow control.

Please note: the Tx and Rx directions are independent byte lanes. Also, when Bypass is enabled the IO paths are automatically disabled.

TxFfIntEn when set enables the Tx direction FIFO Interrupt. The tx FIFO interrupt is latched [available in Status register] and triggered by the Tx FIFO going Almost Empty. If DMA is not used or if smaller DMA transfers are used this interrupt can be useful to trigger a new data transfer in the TX direction.

RxFfIntEn when set enables the Rx direction FIFO Interrupt. The Rx FIFO interrupt is latched [available in Status register] and triggered by the Rx FIFO going Almost Full. If DMA is not used or if smaller DMA transfers are used this interrupt can be useful to trigger a new data transfer in the RX direction.

Please see the status port for more information about the clearing of interrupt conditions.

TxFfAMtIntEnLvI when set enables the level based interrupt for the Tx FIFO Almost Empty condition. Since it is level controlled, the interrupt is cleared by adding more data to the FIFO or by disabling this enable. Since this is the data lane FIFO the write will be through the associated Bus Controller channel. Defined for future implementation or client use. Unused in reference design.

RxFfAFIIntEnLvI when set enables the level based interrupt for the Rx FIFO Almost Full condition. Since it is level controlled, the interrupt is cleared by adding reading data from the FIFO or by disabling this enable. Since this is the data lane FIFO the read will be through the associated Bus Controller channel. Defined for future implementation or client use. Unused in reference design.

TxIoEn when set causes the IO associated with the channel to be configured for



transmission. In the case of the reference design with the bit set the directions are set to cause transmission and the terminations disabled. The default configuration is disabled.

RxIoEn when set causes the IO associated with the channel to be routed from the Receiver. When cleared the data path into the Rx FIFO is from the GPB. Set for standard IO operation, clear for Target writes to the Rx FIFO.

Channel Status PORT		
DATA BIT	DESCRIPTION	
31	IntStat	
30	LocInt	
29-12	spare	
11	UBFfIntLat	
10	BUFfIntLat	
9	ʻO'	
8	ʻO'	
7	UBFfFI	
6	UBFfAfl	
5	UBFfAMt	
4	UBFfMt	
3	BUFfFI;	
2	BUFfAFI	
1	BUFfAMt	
0	BUFfMt	

#### PcieSpartanVI\_USER\_CH\_STATUS

[0x004 Status Register Port read/write]

Figure 30

PCIe-Spartan-VI Channel Status Register

IntStat when set indicates that LocInt is set and the Mask is also enabled i.e. an interrupt is active for this channel.

LocInt is the combination of interrupts and single level masks before the channel level mask. Set when at least one Interrupt request is present.

**note:** UB = User => Bus Controller BU = Bus Controller => User

UBFfIntLat is latched and held until cleared by write to the status port with this bit position set. Almost Full User to Bus Controller FIFO.

BUFfIntLat is latched and held until cleared by write to the status port with this bit



position set. Almost Empty Bus to User FIFO within User.

UBFfFI is set when the User to Bus Controller FIFO is Full. UBFfAFI is set when the User to Bus Controller FIFO is Almost Full. UBFfAMt is set when the User to Bus Controller FIFO is Almost Empty. UBFfMt is set when the UB FIFO is Empty.

BUFfFI is set when the Bus Controller to User FIFO is Full. BUFfAFI is set when the Bus Controller to User FIFO is Almost Full. BUFfAMt is set when the Bus Controller to User FIFO is Almost Empty. BUFfMt is set when the Bus Controller to User FIFO is Empty.

#### PcieSpartanVI\_USER\_CH\_SP

[0x008 Status Register Port read/write]

	Channel Spare Port
<b>DATA BIT</b>	DESCRIPTION
31 - 0	Spare Register

Figure 31

PCIe-Spartan-VI Channel Spare Register

It is always nice to have a spare register to R/W with and make sure your SW is operating properly. Each channel has one so you can do independence testing etc. Store values or ignore in your operational implementation.

### PcieSpartanVI\_USER\_BUS\_FIFO\_CNT

[0x00C Status Register Port read]

Channel UB FIFO Data Count		
<b>DATA BIT</b> 31-0	<b>DESCRIPTION</b> Data positions currently used in UB FIFO	

Figure 32

PCIe-Spartan-VI Channel UB FIFO Count

Bit positions 9-0 have the count. 31-10 are set to 0x00. The count is the 32 bit data count within the User to Bus Controller FIFO. Data is written as LW and read out as bytes before being transferred to the Bus Controller on the corresponding data lane. This count is used to compare against the programmed levels to determine Almost Full etc.



### PcieSpartanVI\_BUS\_USER\_FIFO\_CNT

[0x010 Status Register Port read]

#### Channel BU FIFO Data Count

**DATA BIT** 31-0

**DESCRIPTION** Data positions currently used in BU FIFO

Figure 33

PCIe-Spartan-VI Channel BU FIFO Count

Bit positions 9-0 have the count. 31-10 are set to 0x00. The count is the 32 bit data count within the BU FIFO. Data is written as bytes and read out as LW after being transferred from the Xilinx on the corresponding data lane. This count is used to compare against the programmed levels to determine Almost Empty etc.

The UB and BU FIFOs use flow control which is fixed in HW. Additional programmable [user] level control is provided to support interrupts and other purposes. The data is moved from one FPGA to the other by reading from the upstream port and writing to the downstream port. Since the data transfer is pipelined, there are delays with the status making it back to the controlling port. This is overcome by using burst mode transfers when the transmitting side is not Almost Empty and the Receive side is not Almost Full. When either side is Almost Empty /Full the transfer moves to a slower mode where the status can be checked after each byte is moved to allow the last bytes to be transferred without over-run or under-run issues.

#### PcieSpartanVI\_USER\_FIFO\_WR

[0x014 UB FIFO Write Port]

#### Channel UB FIFO Write

DATA BITDESCRIPTION31-0Data to write to UB FIFO

Figure 34

PCIe-Spartan-VI Channel UB FIFO Write

Writing to this port will put data into the UB FIFO. The data will transfer via the byte lanes through the Bus Controller and into user memory [if DMA is set-up etc.]. The reference software has a loop-back test where data is written to this port and DMA'd back to user space. Useful for debugging initial DMA into system memory without needing DMA out of system memory. Not normally used for system operation.



### PcieSpartanVI\_USER\_FIFO\_RD

[0x014 BU FIFO Read Port]

#### Channel BU FIFO Read

**DATA BIT** 31-0 **DESCRIPTION** Data read from BU FIFO

Figure 35

PCIe-Spartan-VI Channel BU FIFO Read

Reading from this port will retrieve data from the BU FIFO. The data from the Bus Controller side will transfer via the byte lanes from user memory [if DMA is set-up etc.] and into the User BU FIFO. The reference software incorporates a loop-back test where data is DMA'd to the channel's BU FIFO and read back with target accesses from this port. Useful for debugging initial DMA from system memory without needing DMA into system memory. Not normally used for system operation.



## **User FPGA Reference Design**

The FPGA pin definitions are contained in the reference design. See the .UCF file for the project. The IO definitions are the same as PCI-Altera and PCIe-AlteraCycloneIV designs to make porting to this version easy to do.

The reference design is hierarchical and mostly ported from the Altera versions of the design. When porting your design from the Altera to the Xilinx FPGA you will need to update any Altera specific implementations – clock buffers etc. The VHDL/Verilog portion will transfer directly. We have made some improvements and recommend keeping the top of the hierarchy replacing the IO portions of the design. Slightly more typing but the decoding etc. ends up with the same type of controls and should be easy to do.

For new designs we definitely recommend using the reference design and building on top of that. The reference SW package includes a generic R/W instruction allowing the designer to completely redo the memory map and still control the User FPGA features.

The top module defines the connection to the UCF file and the underlying logic. Not strictly necessary – done for consistency with our other designs which frequently have cores or other logic tied in at the top level.

UserAp is the base level of the User Design. The decoding for the registers, clock generation [PLLs etc.] and ports are tied in at this level.

Each of the Ports is tied into the UserAp and provide independent controllers to provide the loop-back function. The FIFO and Control bus interfaces are distributed to provide a FIFO interface within each port. The idea is for the User design to replace the state-machines in the reference design to implement whatever protocol is required.

The IO are evenly distributed in the reference design. The allocation can be updated by the tie in at the UserAp level plus the logic within the ports. The number of ports can be reduced or potentially expanded depending on how the FIFO lanes are operated.



## **Shunts & Headers**

## HDR1 Bus TP

Connected to the dipswitch and used for development purposes [Bus FPGA] not installed for client HW.

## HDR2 User TP

Connected to User FPGA to create access to internal logic. Suitable for scope or logic analyzer.

### H3 User FPGA Flash

User FPGA header for programming QSPI. Pins are labeled in silk-screen.

- 1. 3.3V
- 2. GND
- 3. TCK
- 4. TDO
- 5. TDI
- 6. TMS

Use this header with Impact or similar to use JTAG to connect to the User FPGA. Program the User FPGA with a BIT file or the QSPI with MCS.

Note: a mux is present to allow the JTAG connections to be controlled by the Bus FPGA. If this selection is made the header programming will not be available until deselected or rebooted. It is recommended to use the BIT file until the program is mostly working before programming the QSPI to save time, programming cycles on the FLASH, and allow reversion to the "saved" program until ready to update the QSPI version.

## H2 Connector & Cable Reference

Bezel and D100 shield grounding options. Position 1-2 selects AC coupled [.1 $\mu$ F 50V to ground], 2-3 = DC [direct to ground] and open is unreferenced locally.

## **J1 Cable Power Option**

Select Fused power voltage level. 1-2 = 5V. 2-3 = 3.3V open = unreferenced. Fused power supplied to central pins on D100 connector [see pinout table]. Fuse rated for 1.1A. Self-healing.



## J3 TTL Reference

Select TTL IO pull-up reference. 1-2 = 5V, 2-3 = 3.3V. Open is unreferenced for systems with external pull-ups. Standard pull-ups are 1K ohms. LVC125 single gate devices used for transmitter and receiver.

## LEDs

## L1P8V

Illuminated for 1.8V power supply within tolerance. Used for Bridge.

## L1P2V

Illuminated for 1.2V power supply within tolerance. Used for both FPGAs.

## L5V

Illuminated for 5V power supply within tolerance. Used to generate 2.5V for both FPGAs. Intermediate voltage generated from 12V to guarantee proper voltage sequencing at FPGAs.

## L12V

Illuminated for 12V power supply within tolerance. This is the main reference supplied by the PC. If not on or steady investigate the 12V before proceeding.

### LED3-0

User LEDs oriented to create nibble hex numbers when used. Tied to the User FPGA. For user purposes.



## Loop-Back

Loop-back can be accomplished with the use of an HDEterm100 and HDEcabl100. The following cross connections are used to support the TTL and differential loop-back tests supplied with the driver and UserAp software.

TTL signal "from"	"to"
(11)100	(10)50
(9)44	(8)38
(7)82	(6)32
(5)74	(4)24
(3)68	(2)18
(1)12	(0)6

Each block of IO represents Channels in UserAp software (i.e. IO0-IO4 = Channel 0, IO5-IO9 = Channel 1)

Differential "from"	"to"
<u>(</u> IO0)1,51	(IO5)7,57
(IO1)2,52	(IO6)8,58
<u>(</u> IO2)3,53	(IO7)9,59
(IO3)4,54	(IO8)10,60
(IO4)5,55	(IO9)11,61
<u>(</u> IO10)13,63	(IO15)19,69
(IO11)14,64	(IO16)20,70
<u>(</u> IO12)15,65	(IO17)21,71
(IO13)16,66	(IO18)22,72
(IO14)17,67	(IO19)23,73
<u>(</u> IO20)27,77	(IO25)33,83
(IO21)28,78	(IO26)34,84
<u>(</u> IO22)29,79	(IO27)35,85
(IO23)30,80	(IO28)36,86
(IO24)31,81	(IO29)37,87
<u>(</u> IO30)39,89	(IO35)45,95
(IO31)40,90	(IO36)46,96
<u>(</u> IO32)41,91	(IO37)47,97
(IO33)42,92	(IO38)48,98
(IO34)43,93	(IO39)49,99



## **D100 Standard Pin Assignment**

The pin assignment for the PCIe-Spartan-VI P1 connector.

The pin deelgrink			
IO_0P	IO_0M	1	51
IO_1P	IO_1M	2	52
IO_2P	IO_2M	3	53
IO_3P	IO_3M	4	54
IO_4P	IO_4M	5	55
TTL_0	GND*	6	56
IO_5P	IO_5M	7	57
IO_6P	IO_6M	8	58
IO_7P	IO_7M	9	59
IO_8P	IO_8M	10	60
IO_9P	IO_9M	11	61
TTL_1	GND*	12	62
IO_10P	IO_10M	15	63
IO_11P	IO_11M	14	64
IO_12P	IO_12M	15	65
IO_13P	IO_13M	16	66
IO_14P	IO_14M	17	67
TTL_2	TTL_8	18	68
IO_15P	IO_15M	19	69
IO_16P	IO_16M	20	70
IO_17P	IO_17M	21	71
IO_18P	IO_18M	22	72
IO_19P	IO_19M	23	73
TTL_3	TTL_9	24	74
fused fused VCCB	112_5	25	75
fused fused VCCB		26	76
	10.0014		
IO_20P	IO_20M	27	77
IO_21P	IO_21M	28	78
IO_22P	IO_22M	29	79
IO_23P	IO_23M	30	80
IO_24P	IO_24M	31	81
TTL 4	TTL_10	32	82
IO 25P	IO 25M	33	83
	—		
IO_26P	IO_26M	34	84
IO_27P	IO_27M	35	85
IO_28P	IO_28M	36	86
IO_29P	IO_29M	37	87
TTL_5	GND*	38	88
IO 30P	IO 30M	39	89
IO_31P	IO_31M	40	90
IO_32P	IO_32M	41	91
	—		
IO_33P	IO_33M	42	92
IO_34P	IO_34M	43	93
TTL_6	GND*	44	94
IO_35P	IO_35M	45	95
IO_36P	IO_36M	46	96
IO 37P	IO_37M	47	97
IO_38P	IO_38M	48	98
IO_39P	IO_39M	49	99
TTL_7	TTL_11	50	100

Figure 36

PCIe-Spartan-VI D100 Pinout

VCCB is shunt selectable – 3.3V or 5V. Fused with 1.1A resettable fuse.



## **Applications Guide**

### Interfacing

Some general interfacing guidelines are presented below. Do not hesitate to contact the factory if you need more assistance.

### ESD

Proper ESD handling procedures must be followed when handling PCIe-Spartan-VI. The card is shipped in an anti-static, shielded bag. The card should remain in the bag until ready for use. When installing the card, the installer must be properly grounded and the hardware should be on an anti-static work-station.

#### Start-up

Make sure that the "system" can see your hardware before trying to access it. Many BIOS will display the PCI devices found at boot up on a "splash screen" with the VendorID and CardId and an interrupt level. Look quickly! If the information is not available from the BIOS then a third party PCI device cataloging tool will be helpful. In Windows systems the device manager can be used.

**Watch the system grounds**. All electrically connected equipment should have a failsafe common ground that is large enough to handle all current loads without affecting noise immunity. Power supplies and power consuming loads should all have their own ground wires back to a common point.



## **Construction and Reliability**

PCIe Modules while commercial in nature can be conceived and engineered for rugged industrial environments. PCIe-Spartan-VI is constructed out of 0.062 inch thick High Temp FR4 material.

Surface mount components are used. Most devices are high pin count compared to mass of the device. For high vibration environments inductors and other higher mass per joint components can be glued down.

Conformal Coating is an option. For condensing environments conformal coating is required.

ROHS processing is an option. Standard lead solder is used unless "-ROHS" is added to the part number.

The D100 connector has Phosphor Bronze pins with Nickel plating for durability and Gold plating on the contact area on both plugs and receptacles. The connectors are keyed and shrouded. The pins are rated at 1 Amp per pin, 500 insertion cycles minimum [at a rate of 800 per hour maximum]. These connectors make consistent, correct insertion easy and reliable.

## **Thermal Considerations**

PCIe-Spartan-VI is designed with CMOS circuits. The power dissipation due to internal circuitry is very low. With the one-degree differential temperature to the solder side of the board external cooling is easily accomplished.



## Warranty and Repair

Please refer to the warranty page on our website for the current warranty offered and options.

https://www.dyneng.com/warranty.html

## **Service Policy**

Before returning a product for repair, verify as well as possible that the suspected unit is at fault. Then call the Customer Service Department for a RETURN MATERIAL AUTHORIZATION (RMA) number. Carefully package the unit, in the original shipping carton if this is available, and ship prepaid and insured with the RMA number clearly written on the outside of the package. Include a return address and the telephone number of a technical contact. For out-of-warranty repairs, a purchase order for repair charges must accompany the return. Dynamic Engineering will not be responsible for damages due to improper packaging of returned items. For service on Dynamic Engineering Products not purchased directly from Dynamic Engineering contact your reseller. Products returned to Dynamic Engineering for repair by other than the original customer will be treated as out-of-warranty.

## **Out of Warranty Repairs**

Out of warranty repairs will be billed on a material and labor basis. Customer approval will be obtained before repairing any item if the repair charges will exceed one half of the quantity one list price for that unit. Return transportation and insurance will be billed as part of the repair and is in addition to the minimum charge.

### For Service Contact:

Customer Service Department Dynamic Engineering 150 DuBois, Suite B&C Santa Cruz, CA 95060 (831) 457-8891

support@dyneng.com



# Specifications

PCIe Interfaces:	PCIe 4 lane interface
Access types:	Configuration and Memory space utilized
CLK rates supported:	133.33 MHz oscillator on User FPGA, 8 PLLs to provide programmable frequencies (24). 50 MHz GPB clock, internal DCMs.
Memory	FIFO memory is provided to support DMA both within the Controller and User FPGAs. Flow control is also implemented between the devices. 16 data lanes coupled to make 8 bidirectional full duplex channels.
Ю	40 RS-485 or LVDS transceivers with programmable direction and termination - Twelve TTL with programmable direction.
Interface:	D100 connector. [AMP] 5787082-9 is the board side part number
Software Interface:	Control Registers within Controller and User FPGAs. VHDL reference design for User. Drivers provide generic calls for GPB access to allow any user modification to be programmed with the standard driver.
Initialization:	Programming procedure documented in this manual
Access Modes:	Registers on longword boundary. Standard target access read and write to registers and memory. DMA access to memory.
Access Time:	no wait states in DMA modes. 1-2 wait states in target access to Bus FPGA.
Interrupt:	1 interrupt is supported with multiple sources. The interrupts are maskable and are supported with a status register.
Onboard Options:	Shunt for 3.3V or 5V reference to TTL IO. AC/ DC reference for cable shield. 3.3V /5V fused power at connector. Header position for direct programming of User FPGA or attached QSPI FLASH.
Dimensions:	half-length PCIe board.
Construction:	High Temp FR4 Multi-Layer Printed Circuit, Surface Mount Components.
Power:	12V, 3.3V from PCIe bus. Local 5V and 2.5V, 1.8V, and 1.2V created with on-board power supplies.
User	8 position software readable switch 4 software controllable LEDs Power Supply LEDs



## **Order Information**

Standard temperature range -40-85°C components PCIe-Spartan-VI https://www.dyneng.com/PCIe-Spartan-VI.html half length PCIe card with user re-configurable Spartan VI [XC6S100LX-3FG676I is base device], 40 differential IO, 12 TTL IO, 8 PLL [24 clocks] On-The-Fly programming for the User FPGA. QSPI for instant on also included. PCIe-Spartan-VI-ENG Engineering Kit for the PCI-Spartan-VI is ala-carte Software, Cable, HDEterm100, reference User implementation. Please refer to the webpage for more information. Driver / reference SW: Win10/11, Linux. -LVDS Switch to LVDS IO instead of RS485(default selection). -ROHS Change to ROHS processing. Otherwise, leaded solder is used. -CC Add Conformal Coating for condensing environments -Mixed "Mixed" will be replaced with a new identifier. Combination of LVDS and RS485 IO installed. HDEterm100 https://www.dyneng.com/HDEterm100.html 100-pin connectors (2) matching the PCIeAlteraCycloneIV-485/LVDS D100 interconnected with 100 screw terminals. DIN rail mounting. Optional terminations and test points. HDEcable100 https://www.dyneng.com/HDEcabl100.html 100 pin connector matching PCIe-Spartan-VI and HDEterm100. Length options

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## Appendix

	READ ADDRESS		WRITE ADDRESS	
CNTL_REN				
READx				
CNTL_DATA	READ DATA		WRITE DATA	
CNTL_WEN	Data Latche	ed In Xilinx		
LOADx				

### **General Purpose Bus Timing**

The basic timing for the GPB relative to the User device is shown. The CNTL clock operates at 50 MHz. Signals originating from the Bus Controller are registered to provide set-up and hold. Skew is kept to a minimum with edge FF's and controlled routing within the PCB.

A host read of the User over the GPB starts with the address and Read Enable being asserted. The User logic decodes the address and CNTL\_REN to generate the local READx signal. The reference design provides 180 of the READ decodes. Read data is presented back to the CNTL\_DATA bus and captured coincident with the edge of CNTL\_REN as shown.

A host write to the User over the GPB starts with the address and data being asserted. After a 1 period delay CNTL\_WEN is asserted. The User logic decodes the Address and CNTL\_WEN to create the LOADx signals. LOADx when used as a write enable allows the CNTL\_DATA to be loaded into the target device.

The included reference design [VHDL] provides the decoding and sample registers.

Further documentation is provided in-line with the VHDL.

